



## Analysis of Light-Controlled Microstrip Filter

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# Analysis of Light-Controlled Microstrip Filter

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In this paper, we propose light-controlled microstrip filters whose frequency band can be changed by light illumination which induces plasma in a silicon chip buried in part of a substrate. We analyze the frequency characteristics of the filters by using the (FD)<sup>2</sup>TD method and investigate the characteristics affected by the photo-excited plasma.

## 1. Introduction

Recently, high-resistivity semiconductors have been expected to be used as substrate materials for microwave integrated circuits<sup>1-4</sup>. The illuminating of light whose photon energy is larger than the bandgap energy of the semiconductor induces electron-hole plasma in a light-illuminated area. The small region filled with photo-excited electron-hole plasma, as distinct from the surrounding region, has a complex permittivity and hence modulates microwaves propagating through<sup>5-7</sup>. Such optical control offers the following advantages: a) nearly perfect isolation, b) low static and dynamic insertion losses in some regimes, c) fast response, d) capability of high power handling and so on. In this study, we propose light-controlled microstrip filters whose frequency band can be changed by light illumination.

One of numerical schemes which have been successfully used to calculate the time-domain fields is the FDTD (Finite Difference Time Domain) method. It was first proposed by K.S.Yee and has been applied by many investigators to the solution of various electromagnetic problems. The FDTD method shows great promise in its flexibility in handling a wide variety of circuit configurations<sup>8-12</sup>. Broad-band pulses, if used as the initial condition for excitation, enable us to calculate the frequency-domain parameters over the entire frequency range of interest. Recently, the FDTD method has been extended to deal with dispersive

materials like plasma, called the (FD)<sup>2</sup>TD (Frequency Dependent Finite Difference Time Domain) method.

In this paper, we analyze frequency characteristics of light-controlled microstrip filters by using the (FD)<sup>2</sup>TD method and investigate the effects of photo-excited plasma region on the frequency characteristics.

## 2. Analysis

The configuration of the light-controlled microstrip filter is shown in Fig.1. A thin rectangle of silicon chip 1.2mm long and 1.2mm wide is buried into the dielectric substrate on one end of the center microstrip conductor as shown in Fig.1(model 1). Light beam illuminates the chip on the region  $L_1$  [mm]  $\times$  1.2mm, inducing photo-excited plasma. In this analysis, the equipment of light beam emission isn't taken into consideration for the convenience of analysis. By changing the length  $L_1$ , frequency band of the filter is controlled. The parameters of the structure used in numerical calculations are as follows:

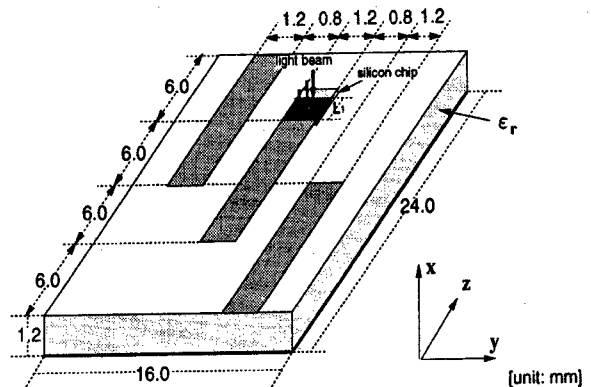


Fig.1 Configuration of the light-controlled microstrip filter (model 1)

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width of the microstrip conductor: 1.2mm,  
 thickness of the substrate: 1.2mm  
 dielectric constant of the substrate:  $\epsilon_r=4.4$ ,  
 length of the center microstrip conductors:  
 12mm,  
 space between microstrip conductors: 0.8mm.

The relative complex permittivity of the semiconductor is given by

$$\epsilon_r = \epsilon_s - \sum_{i=e,h} \left[ \frac{\omega_{pi}^2}{\omega^2 + \nu_i^2} + j \frac{\nu_i}{\omega} \frac{\omega_{pi}^2}{\omega^2 + \nu_i^2} \right] \quad (1)$$

$$\omega_{pi} = \left[ \frac{-ne^2}{\epsilon_0 m_i} \right]^{\frac{1}{2}} \quad (i=e, h)$$

where  $\omega$  is the angular frequency of the electromagnetic fields,  $e$  ( $=1.60 \times 10^{-19}$  kg) the elementary charge,  $\epsilon_s$  ( $=11.8$ ) the permittivity of the semiconductor without plasma,  $m_e$  ( $=0.259m_0$ ) the effective mass of an electron,  $m_h$  ( $=0.38m_0$ ) the effective mass of a hole,  $m_0$  ( $=9.11 \times 10^{-31}$  kg) the rest mass of an electron,  $\nu_e$  ( $=4.53 \times 10^{12}$  s $^{-1}$ ) the collision angular frequency for electrons and  $\nu_h$  ( $=7.31 \times 10^{12}$  s $^{-1}$ ) the collision angular frequency for holes.

In the analysis, we set up uniform grids with a difference  $\Delta=0.100$  mm in space. Three dimensions of the whole structure are  $40\Delta \times 160\Delta \times 240\Delta$ . We excite the input port at  $z=12\Delta$  with a Gaussian pulse in time. It is uniform on the transverse plane under the microstrip conductor and  $E_x$  is the only nonvanishing component of electric fields, expressed as

$$E_x(t) = -\frac{2(t-t_0)}{T^2} \exp\left\{-\frac{(t-t_0)^2}{T^2}\right\} \quad (3)$$

where  $t_0=200\Delta t$  and  $T=60\Delta t$ .

Scattering parameters  $S_{11}$  and  $S_{21}$  are given by

$$S_{11}(f) = \frac{V_{ref}(f, z_1) - V_{inc}(f, z_1)}{V_{inc}(f, z_1)} \quad (4)$$

and

$$S_{21}(f) = \frac{V_{trs}(f, z_2)}{V_{inc}(f, z_1)} \quad (5)$$

respectively, where  $V_{ref}$  is the Fourier transform of the voltage of reflected waves at the input port,  $V_{inc}$  the

Fourier transform of the voltage of incident waves at the same position, and  $V_{trs}$  the Fourier transform of the voltage of transmitted waves at the output port.

Figure 2 shows scattering parameters when  $L_1=0.4$  mm and 1.2 mm in model 1. This result indicates that the center frequency of pass band is shifted from 5.92 GHz to 5.59 GHz by increasing the light-illuminated area from  $0.4 \times 1.2$  mm $^2$  ( $L_1=0.4$  mm) to  $1.2 \times 1.2$  mm $^2$  ( $L_1=1.2$  mm). On the other hand, the insertion loss increases from 2.78 dB ( $L_1=0.4$  mm) to 3.68 dB ( $L_1=1.2$  mm).

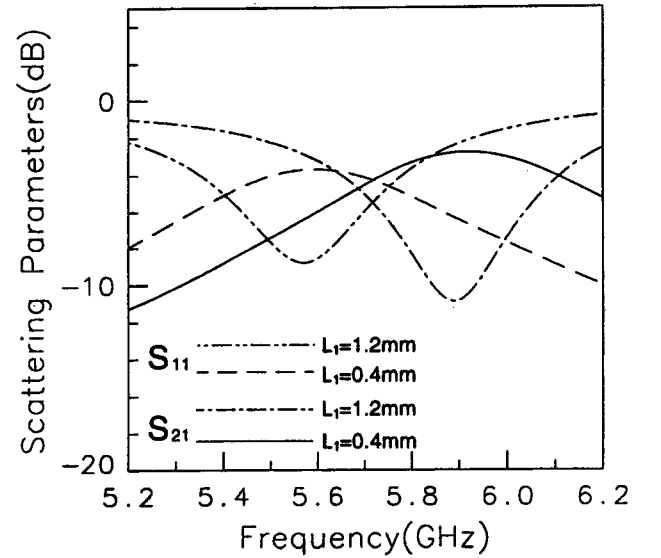


Fig.2 Scattering parameters (model 1)

Next, we consider another type of light-controlled filter to suppress insertion losses. Two silicon chips 0.6 mm long and 1.2 mm wide are buried into the dielectric substrate on both ends of the center microstrip conductor as shown in Fig.3 (model 2). Light

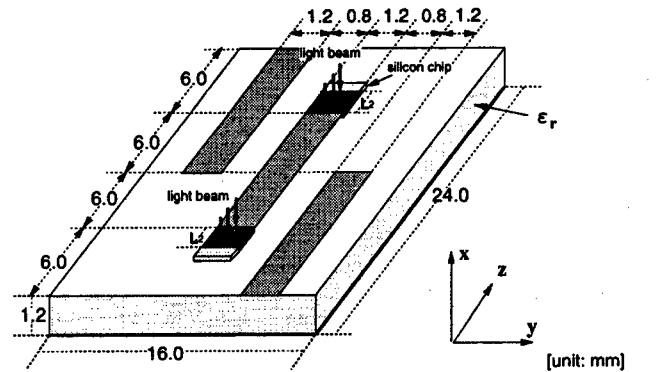


Fig.3 Configuration of the light-controlled microstrip filter (model 2)

beam illuminates the chips simultaneously on the region of  $L_2$  [mm]  $\times$  1.2mm.

Figure 4 shows scattering parameters when  $L_2 = 0.2$  mm and 0.6mm in model 2. It is found that the center frequency of pass band is shifted from 5.81 GHz to 5.42GHz by increasing the light-illuminated area from  $0.2 \times 1.2$ mm<sup>2</sup> ( $L_2 = 0.2$ mm) to  $0.6 \times 1.2$  mm<sup>2</sup> ( $L_2 = 0.6$ mm). Although the total area illuminated by light is quite the same as model 1, the insertion loss is 2.90 dB when  $L_2 = 0.2$  mm and 3.55 dB when  $L_2 = 0.6$ mm; hence the insertion loss is suppressed for the filter with a wider light-illuminated region.

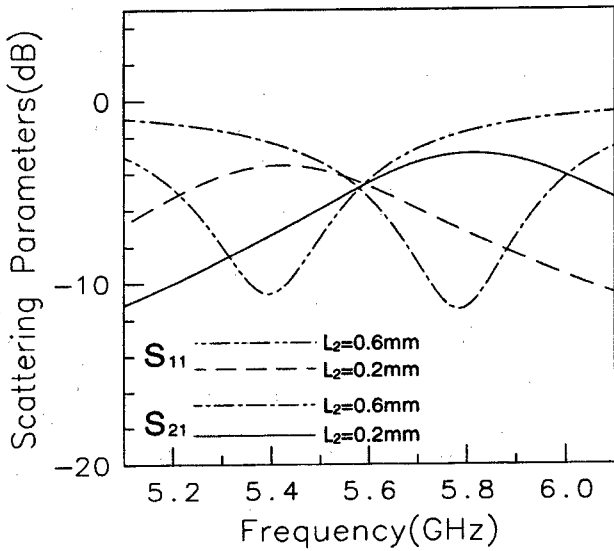


Fig.4 Scattering parameters (model 2)

Finally, we investigate the effect of the photo-excited plasma region to give a reasonable explanation. Figure 5 shows a microstrip open end followed by the plasma region of  $L_3$  [mm]  $\times$  1.2mm. We calculate the conductance of the plasma region seen at the end of

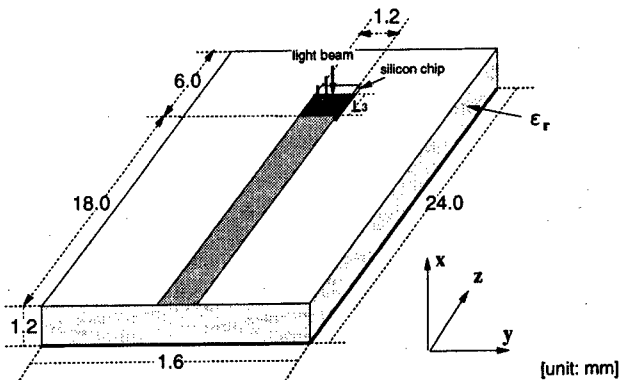


Fig.5 Configuration of the microstrip open-end which has the plasma region

microstrip. The conductance  $G(f)$  is determined in terms of  $S_{11}(f)$  as follows:

$$G(f) = Re \left[ Z_0(f) \cdot \frac{1 + S_{11}(f)}{1 - S_{11}(f)} \right] \quad (6)$$

In this equation,  $Re[]$  means to take the real part of the factor in the brackets and  $Z_0(f)$  is a characteristic impedance of the infinitely long microstrip.

Figure 6 shows the conductance  $G(f)$  versus frequency for  $L_3 = 0.2$ mm with the solid line and for 0.4mm with the dashed line in (a) and for  $L_3 = 0.6$ mm with the solid line and for 1.2 mm with the dashed line in (b). We double the data shown with the solid lines, drawing the resultant values with the dash-dotted lines in both figures. Because of the symmetry in structure, the conductance of model 2 can be estimated by the dash-dotted line. As understood from the figure, the conductance of model 1 is evaluated to be smaller than model 2 when  $L_1 = 0.4$ mm in model 1 and  $L_2 = 0.2$  mm in model 2, and vice versa when  $L_1 = 1.2$  mm in model 1 and  $L_2 = 0.6$  mm in model 2. These results

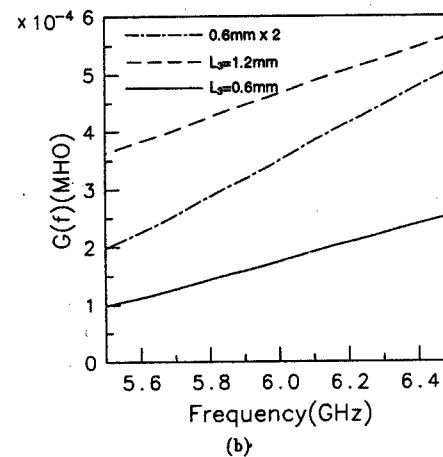
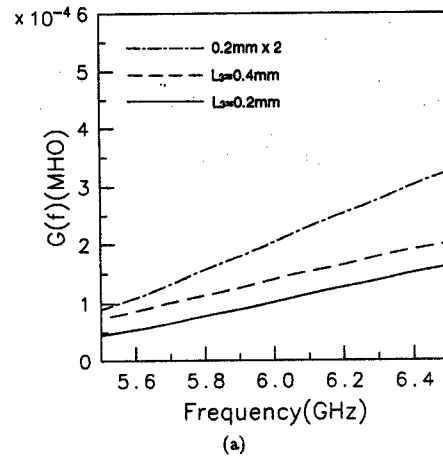


Fig.6 Frequency characteristics of conductance

suggest that the conductance is rather on the exponential increase as the length of light-illuminated region becomes longer.

### 3 Conclusion

In this paper, we have proposed light-controlled microstrip filters. We have analyzed the frequency characteristics by using the (FD)<sup>2</sup>TD method and demonstrated that the frequency band can be controlled by changing the light-illuminated area. We also have investigated insertion losses effected by the volume of photo-excited plasma.

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